Review Doppler Radar (Fig. 3.1) A simplified block diagram





Complex plane (Phasor diagram) Electric field incident on scatterer

Reflected electric field incident on antenna

Voltage input to the synchronous detectors; This pair of detectors shifts the frequency *f* to 0

$$V_{o} = I + jQ \propto A_{o} \exp\left[-j4\pi r / \lambda + j\psi_{t}\right]$$

Echo voltage V_o at the output of the detectors and filters . I(t) The echo amplitude is A_o and phase is

$$\psi_{\rm e} = (-4\pi r \,/\,\lambda) + \psi_t$$

If the range *r* of the scatterer is fixed, the phasor (A_o, ψ_e) is fixed (i.e., no change in A_o nor ψ_e . But if scatterer has a radial velocity, phasor (A_o, ψ_e) rotates about the origin at the Doppler frequency f_d .



Pulsed Radar Principle





- c = speed of microwaves
 - $= c_h \text{ for H and } = c_v \text{ for V waves}$
- τ = pulse length
- λ = wavelength
- $= \lambda_{\rm h}$ for H and $\lambda_{\rm v}$ for V waves
- $\tau_s =$ time delay between

transmission of a pulse and reception of an echo.

Angular Beam Formation

(the transition from a circular beam of constant diameter to an angular beam of constant angular width)



Antenna (directive) Gain g_t The defining equation:

$$S_i = \frac{P_t}{4\pi r^2} g_t f^2(\theta, \phi)$$
 Eq. (3.4)

 S_i (W m⁻²) = Power density incident on a scatterer r = range to measurement (m) $f^2(\theta, \phi)$ = radiation pattern = 1 on beam axis P_t = transmitted power (W)

Backscattering Cross Section, σ_b for a Spherical Particle

Rayleigh condition on a spherical particle of diameter D:

$$D < \lambda / 16; \quad \lambda \equiv \text{wavelength}$$

$$\sigma_{\rm b} = \frac{\pi^5}{\lambda^4} |\mathbf{K}_{\rm m}|^2 \mathbf{D}^6;$$

$$K_m = \frac{m^2 - 1}{m^2 + 2}; \text{ Dielectric Factor of the medium filling the sphere; Eq.3.6}$$

$$m = n - jn\kappa = \text{the complex index of refraction}$$

$$|K_m|^2 \rightarrow |K_w^2 = 0.93 \text{ for water, and}$$

$$|K_m|^2 \rightarrow |K_i|^2 = 0.18 \text{ for ice (density} = 0.917 \text{ g m}^{-3})$$

Backscattered Power Density Incident on
Receiving Antenna

$$S_{i}$$

$$S_{r}(r,\theta,\phi) = \frac{P_{t}g_{t}f^{2}(\theta,\phi)}{4\pi r^{2}l} \bullet \sigma_{b} \bullet \frac{1}{4\pi r^{2}l} \quad (3.13a)$$

where 1 is the loss factor (due to attenuation)

$$1 = \exp\left(\int_{0}^{r} (k_g + k)dr\right)$$

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(3.13b)

Echo Power P_r Received

$$P_r = S_r(r,\theta,\phi)A_e(\theta,\phi) \quad (3.20)$$

 A_e is the effective area of the receiving antenna for radiation from the θ, φ direction. It is shown that:

$$A_e = g_r f_r^2(\theta, \phi) \lambda^2 / 4\pi \qquad (3.21)$$

If the transmitting antenna is the same as the receiving antenna then:

$$g_r f_r^2(\theta, \phi) = g_t f_t^2(\theta, \phi) \equiv g f^2(\theta, \phi)$$

The Radar Equation (point scatterer/discrete object)

$$P_{r} = \frac{P_{t}gf^{2}(\theta,\varphi)}{4\pi r^{2}l} \bullet \frac{\sigma_{b}}{4\pi r^{2}l} \bullet \frac{g\lambda^{2}f^{2}(\theta,\varphi)}{4\pi} \quad (3.24)$$

Example:

 $\lambda = 0.1 \text{ m}; \quad r = 20 \text{ km}(2x10^4 \text{ m}); \quad P_r(\text{min}) = 10^{-14}(\text{ W});$ $P_t = 10^6(\text{ W}; \text{ peak}); \quad g = 3x10^4; \quad l = 1(\text{ no path loss})$ Calculating the minimum detectable backscattering σ_b : $\sigma_b(\text{min}) = 2x10^{-7} \text{ m}^2 = \sigma_b \text{ for a 6.3 mm drop!}$

Unambiguous Range r_a

- If targets are located beyond $r_a = cT_s/2$, their echoes from the n^{th} transmitted pulse are received after the $(n+1)^{\text{th}}$ pulse is transmitted. Thus, they appear to be closer to the radar than they really are!
 - This is known as range folding



• $T_{\rm s} = PRT$

- Unambiguous range: $r_a = cT_s/2$
 - Echoes from scatterers between 0 and r_a are called **1**st trip echoes,
 - Echoes from scatterers between r_a and $2r_a$ are called **2nd trip** echoes, Echoes from scatterers between $2r_a$ and $3r_a$ are called **3rd trip** echoes, etc.

Unambiguous Velocity



- A pulsed Doppler radar measures radial Doppler velocity by <u>keeping track</u> of phase changes between samples that are T_s (pulse repetition time) apart
- Recall that echo phase shift is $\psi_e = -4\pi r/\lambda$. Then, the phase change from pulse to pulse is $\Delta \psi_e = -4\pi \Delta r/\lambda = -4\pi v_r T_s/\lambda$
 - Note that only phase changes between $-\pi$ and π can be unambiguously resolved
- Therefore, the unambiguous velocity is:
 - $4\pi v_a T_s / \lambda = \pi \Rightarrow v_a = \lambda / 4 T_s$
 - This is related to the Nyquist sampling theorem:
 Doppler velocities outside the ±v_a interval will be aliased!

 $\Delta r = v_r T_s$ is the change in range of the scatterer between successive transmitted pulses

Another PRT Trade-Off

• Correlation of pairs: $\rho(T_s) = \exp\left[-8\left(\pi\sigma_v T_s/\lambda\right)^2\right]$

This is a measure of signal coherency

• Accurate measurement of power requires long PRTs

$$\lim_{T_{s}\to\infty}\rho(T_{s})=0$$

– More independent samples (low coherency)

• But accurate measurement of velocity requires short PRTs

$$-\lim_{T_s\to 0}\rho(T_s)=1$$

- High correlation between sample pairs (high coherency)
- Yet a large number of **independent sample pairs** are required

Signal Coherency

- How large a T_s can we pick?
 - Correlation between m = 1 pairs of echo samples is:

$$\rho(T_{s}) = \exp\left[-8\left(\pi \,\sigma_{v} T_{s}/\lambda\right)^{2}\right]$$

Correlated pairs: $\rho(T_{s}) \approx 1 \Rightarrow \frac{\pi \sigma_{v} T_{s}}{\lambda} <<1 \Rightarrow \frac{\lambda}{\pi T_{s}} >> \sigma_{v}$

(i.e., Spectrum width σ_v must be much smaller than unambiguous velocity $v_a = \lambda/4T_s$)

- Increasing T_s decreases correlation exponentially $- \quad Var[\hat{v}] \text{ and } Var[\hat{\sigma}_v] \text{ also increases exponentially!}$
- Pick a threshold:

$$- \rho(T_s) \ge e^{-0.5} \Longrightarrow -8(\pi\sigma_v T_s / \lambda)^2 = 0.5 \Longrightarrow \sigma_v \le V_a / \pi$$

- Violation of this condition results in very large errors of estimates!

Signal Coherency and Ambiguities

- Range and velocity dilemma: $r_a v_a = c\lambda/8$
- Signal coherency: $\sigma_v \leq v_a / \pi$
- r_a constraint: $r_a \le \frac{\sigma \pi}{8\pi \sigma_v}$
 - This is a more basic constraint on radar parameters than the first equation above

Eq. (7.2c)

- Then, σ_v and not v_a imposes a basic limitation on Doppler weather radars
- Example: Severe storms have a median $\sigma_v \sim 4$ m/s and 10% of the time $\sigma_v > 8$ m/s. If we want accurate Doppler estimates 90% of the time with a 10-cm radar ($\lambda = 10$ cm); then, $r_a \leq 150$ km. This will often result in range ambiguities



Echoes (I or Q) from Distributed Scatterers (Fig. 4.1)



Weather Echo Statistics (Fig. 4.4)



Reflectivity Factor Z (Spherical scatterers; Rayleigh condition: $D \le \lambda/16$)

$$\eta(\mathbf{r}) = \frac{\pi^5}{\lambda^4} |K_{\rm m}|^2 Z(\mathbf{r})$$
 (4.31)

where

$$Z(\mathbf{r}) = \frac{1}{\Delta V} \sum_{i} D_{i}^{6} = \int_{0}^{\infty} N(D, \mathbf{r}) D^{6} dD \qquad (4.32)$$
$$\eta(\mathbf{r}) = \frac{\pi^{5}}{\lambda^{4}} |K_{w}|^{2} Z_{e}(\mathbf{r}) \qquad (4.33)$$

for water drops : $|K_w|^2 \approx 0.93$ independent of T(°C); for ice particles : $|K_i|^2 \approx 0.16$ dependent on T and ice density.